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# A Comparative Study of Modular Axial Flux Podded Generators for Marine Current Turbines

Sofiane Djebbari<sup>1,2</sup>, Mohamed Benbouzid<sup>2</sup>, Jean Frédéric Charpentier<sup>1</sup> and Franck Scuiller<sup>1</sup>

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**Abstract**—This research note deals with performance comparison of axial flux modular podded generators for marine current turbines (MCTs). Due to the submarine environment, maintenance operations are very hard, very costly, and strongly depending on sea conditions. In this context, the drive train reliability is a key feature for MCTs. For that purpose, a comparative study is proposed, to assess modular axial flux permanent magnet (AFPM) machines potential for reliability improvement. Thereby, designs of direct-drive modular AFPM generators for a given experimental MCT are performed. The proposed study shows that a pair number of spatially shifted AFPM machine modules, adequately associated, leads to the reduction of the electromagnetic torque ripples transmitted to the MCT shaft. Moreover, it is shown that the proposed module-based generator configuration achieves better thermal behavior. As the active parts masses and costs are expected to be higher, compromises should be carried-out in terms of reliability and fault-tolerance.

**Keywords:** Marine current turbine, axial flux permanent magnet generator, design, optimization.

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## Nomenclature

MCT = Marine Current Turbine;  
AFPM = Axial Flux Permanent Magnet.

## I. Introduction

Marine energy has become an issue of significant interest achieving a spectacular increase in the last years. It is currently the focus of much industrial and academic research around the world [1-2]. Indeed, the astronomic nature of this resource makes it predictable, to within 98% accuracy for decades, and independent of prevailing weather conditions. This predictability is critical to a successful integration of renewable energy in the electrical grid. Nevertheless, several marine energy projects over the world are facing difficulties delaying their complete achievement. These difficulties mainly concern installations high-cost and maintenance [3]. Marine current turbines are similar in many aspects to wind turbine technologies. However, because of tide low speeds and to avoid blade cavitations, the turbine rotational speed is typically below 50rpm. For conventional industrial generators, the rated speed is typically between 1000 and 3000 rpm. The use of multistage gearboxes is therefore needed. Such gearboxes lead to low drive train efficiency and high maintenance requirements. To make tidal current energy conversion economically interesting, Marine Current Turbines (MCTs) will need to have an approximately 30 year lifespan with maintenance inspections every 5 years [4]. Therefore, MCTs should be highly efficient and reliable.

Direct-drive permanent magnet generators appear as a solution that can fulfill these specific requirements [5-6].

However, the generator active parts mass and cost are expected to be higher if compared with more conventional high speed industrial generators [7]. In other hand, direct-drive permanent magnets generators are characterized by high torque ripples (around 10% of the rated electromagnetic torque); these ripples can be reduced using a fractional winding distribution. However, this makes the winding manufacturing harder.

In this research note, the focus is on the optimal design and the spatial arrangement of modular AFPM generators in order to reduce the electromagnetic torque ripples (Fig. 1) [8]. Respectively, one, two, and four modules are optimally designed and compared in terms of active parts costs and masses, and torque ripple on the turbine shaft. In this context and regarding MCTs design, a POD topology seems more favorable than a rim-driven one to achieve multi-module arrangements. For illustration, Fig. 2 shows some of the relevant podded marine current turbine projects [9-10].

## II. Design Tools and Methodology

### II.1. Design Specifications

The proposed study is based on the specification of a real MCT (300kW) [10-11]. Table 1 gives the specification parameters set used in the design optimization process.

### II.2. AFPM Generator Modeling

The AFPM generator geometry is modeled as its equivalent linear machine developed at mean radius. Figure 3 shows the geometry parameters.

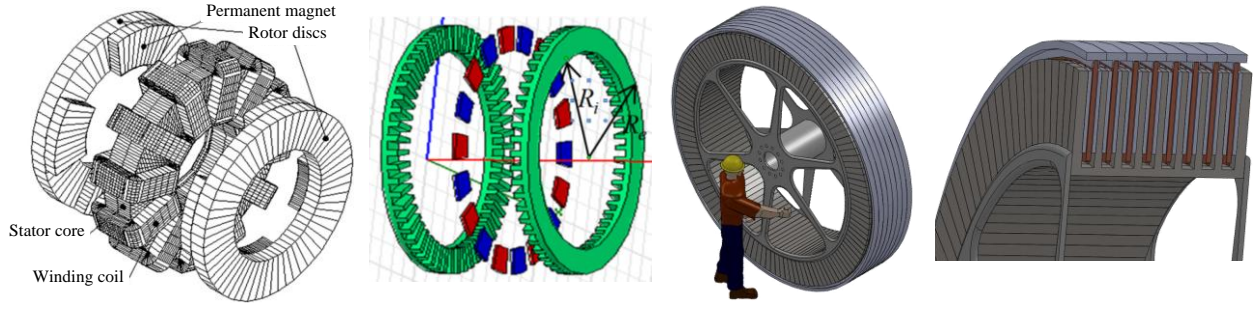


Fig. 1. Examples of axial flux permanent magnet machine concept [8-10] (Illustration reproduced with the kind permission of the authors [8]).



(a) Alstom/TGL turbine [©Alstom].

(b) Voith turbine [©Voith].

(c) Atlantis turbine [©Atlantis].

Fig. 2. Relevant example of podded marine current turbines.

Table 1. DESIGN SPECIFICATION SET.

|  |             |        |     |
|--|-------------|--------|-----|
| Turbine radius (Seaflow)                         | $R_0$       | 5.5    | m   |
| Torque transmitted by the turbine                | $Q$         | 191    | kNm |
| Turbine speed                                    | $N$         | 15     | rpm |
| Magnet to pole width ratio                       | $\beta_m$   | 0.65   | -   |
| Slot fill factor                                 | $k_f$       | 0.65   | -   |
| Machine electrical frequency                     | $f_{mach}$  | 50     | Hz  |
| Electrical angle                                 | $\psi$      | 0      | rad |
| Phases number                                    | $m$         | 3      | -   |
| Slot number per pole per phase                   | $S_{pp}$    | 1      | -   |
| Magnet coercive field                            | $H_{cj}$    | $10^6$ | A/m |
| Magnet remanent flux density                     | $B_r$       | 1.2    | T   |
| Maximum magnetic flux density in the iron sheets | $B_{sat}$   | 1.4    | T   |
| Conductors maximum temperature                   | $T_{max}$   | 100    | °C  |
| Sea water temperature                            | $T_{water}$ | 30     | °C  |

The AFPM generator electromagnetic modeling is based on an analytical solving of Maxwell equations by variable separation. Then the electromagnetic field is calculated in the permanent magnets and the air gap regions by considering the equivalent slotless machine. This machine is developed by adding a carter coefficient as shown in Fig. 4.

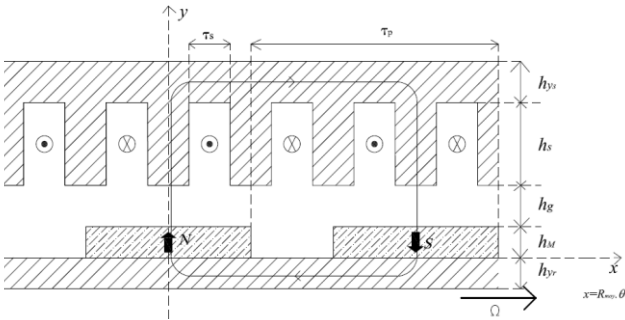


Fig. 3. AFPM generator geometry.

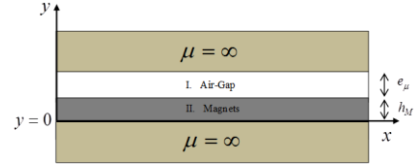


Fig. 4. AFPM generator model representation.

The magnetic field is calculated by solving the following equations [12].

$$\begin{cases} \frac{\partial^2 A_z(x, y)}{\partial x^2} + \frac{\partial^2 A_z(x, y)}{\partial y^2} = 0 \\ \text{in the air gap (region I)} \\ \frac{\partial^2 A_z(x, y)}{\partial x^2} + \frac{\partial^2 A_z(x, y)}{\partial y^2} = -\mu_0 \frac{\partial M_y(x)}{\partial x} \\ \text{in the permanent magnets (region II)} \end{cases} \quad (1)$$

### II.3. Optimization Problem

The optimization objective is to minimize the total cost of the active parts denoted  $C(x)$ .  $C(x)$  is calculated by considering the machine active parts weight described by vector  $x$ . Considering this vector; it is possible to define all the AFPM generator geometry [11]. Relation (2) summarizes the optimization problem.

$$\begin{cases} x^* = \min_{x \in X} \|C(x)\| \\ T_c(x) \leq T_{c \max} \\ H_{c \max}(x) \leq H_{cj} \\ \eta_{elec} \geq \eta_{elec \min} \\ R_e \leq R_{e \max} \end{cases} \quad (2)$$

Where  $x$  is the vector defining the optimization variables:

$$x = [B_{g\max} \quad A_L \quad J \quad p \quad R_e]^T$$

The generator external radius is constrained by  $R_{emax}$  (set to 1/3 of the turbine radius).

### III. Design Results

Table 2 gives the design results of one AFPM optimization. This generator is sized for the 300kW rated power of the turbine. In Table 3, the AFPM generator is optimized for the turbine 1/2 power (150kW). To fulfill the turbine power specifications, two modules containing each one a 150kW AFPM machine are linked to the MCT. In Table 4, the AFPM machine is optimized for the 1/4 rated power of the MCT (75kW). To fulfill the turbine power specification four modules containing each one a 75kW AFPM generator are linked to the MCT.

Figure 5 illustrates the comparison of the total active parts masses and costs, generator maximum temperature, and the torque-to-mass ratio of the different optimized AFPM generators. This figure obviously shows that the AFPM generators modular topology leads to active parts masses and costs oversizing. However for a modular integration, the generator thermal behavior is improved.

In direct-drive permanent magnet machines, the torque ripple is very high (tens of kNm).

Table 2. DESIGN PARAMETERS  
OF A 300kW 1-MODULE AFPM GENERATOR.

| 300kW AFPM Generator           |               |        |                      |
|--------------------------------|---------------|--------|----------------------|
| Current density                | $J$           | 4.5    | A/m <sup>2</sup> rms |
| Electrical load                | $A_L$         | 120000 | A/m rms              |
| Air gap flux density           | $B_{gmax}$    | 0.5    | T                    |
| Pole pairs poles number        | $p$           | 120    | -                    |
| Inner radius                   | $R_i$         | 1.82   | m                    |
| Outer radius                   | $R_e$         | 2      | m                    |
| Generator ring thickness       | $\Delta R$    | 17.91  | cm                   |
| Mean radius                    | $R_m$         | 1.91   | m                    |
| Magnet to pole width ratio     | $\beta_m$     | 66     | %                    |
| Teeth pitch ratio              | $\beta_t$     | 52     | %                    |
| Rotor yoke thickness           | $h_{Yr}$      | 0.86   | cm                   |
| Stator yoke thickness          | $h_{Ys}$      | 0.86   | cm                   |
| Slot height                    | $h_s$         | 8.5    | cm                   |
| Magnets thickness              | $h_M$         | 0.73   | cm                   |
| Air gap (magnet/stator)        | $h_g$         | 0.76   | cm                   |
| Copper maximum temperature     | $T_{cmax}$    | 70     | °C                   |
| Electrical efficiency          | $\eta_{elec}$ | 90     | %                    |
| Magnets maximum magnetic field | $H_{max}$     | 0.62   | MA/m                 |
| Active parts total mass        | $Mass$        | 1840   | kg                   |
| Active parts total cost        | $Cost$        | 11.7   | k€                   |
| Torque/active parts mass       | $T_{EM}/Mass$ | 104    | Nm/kg                |

Table 3. DESIGN PARAMETERS  
OF A 150kW 2-MODULES AFPM GENERATOR.

| 150kW AFPM Generator           |               |        |                      |
|--------------------------------|---------------|--------|----------------------|
| Current density                | $J$           | 3.5    | A/m <sup>2</sup> rms |
| Electrical load                | $A_L$         | 120000 | A/m rms              |
| Air gap flux density           | $B_{gmax}$    | 0.45   | T                    |
| Pole pairs poles number        | $p$           | 144    | -                    |
| Inner radius                   | $R_i$         | 1.79   | m                    |
| Outer radius                   | $R_e$         | 1.9    | m                    |
| Generator ring thickness       | $\Delta R$    | 10     | cm                   |
| Mean radius                    | $R_m$         | 1.84   | m                    |
| Magnet to pole width ratio     | $\beta_m$     | 66     | %                    |
| Teeth pitch ratio              | $\beta_t$     | 46.4   | %                    |
| Rotor yoke thickness           | $h_{Yr}$      | 0.62   | cm                   |
| Stator yoke thickness          | $h_{Ys}$      | 0.62   | cm                   |
| Slot height                    | $h_s$         | 9.8    | cm                   |
| Magnets thickness              | $h_M$         | 0.63   | cm                   |
| Air gap (magnet/stator)        | $h_g$         | 0.74   | cm                   |
| Copper maximum temperature     | $T_{cmax}$    | 68     | °C                   |
| Electrical efficiency          | $\eta_{elec}$ | 90     | %                    |
| Magnets maximum magnetic field | $H_{max}$     | 0.61   | MA/m                 |
| Active parts total mass        | $Mass$        | 1200   | kg                   |
| Active parts total cost        | $Cost$        | 7.25   | k€                   |
| Torque/active parts mass       | $T_{EM}/Mass$ | 80     | Nm/kg                |

Table 4. DESIGN PARAMETERS  
OF A 75kW 4-MODULES AFPM GENERATOR.

| 75kW AFPM Generator            |               |       |                      |
|--------------------------------|---------------|-------|----------------------|
| Current density                | $J$           | 3     | A/m <sup>2</sup> rms |
| Electrical load                | $A_L$         | 80000 | A/m rms              |
| Air gap flux density           | $B_{gmax}$    | 0.45  | T                    |
| Pole pairs poles number        | $p$           | 129   | -                    |
| Inner radius                   | $R_i$         | 1.6   | m                    |
| Outer radius                   | $R_e$         | 1.7   | m                    |
| Generator ring thickness       | $\Delta R$    | 10    | cm                   |
| Mean radius                    | $R_m$         | 1.65  | m                    |
| Magnet to pole width ratio     | $\beta_m$     | 66    | %                    |
| Teeth pitch ratio              | $\beta_t$     | 42.7  | %                    |
| Rotor yoke thickness           | $h_{Yr}$      | 0.57  | cm                   |
| Stator yoke thickness          | $h_{Ys}$      | 0.57  | cm                   |
| Slot height                    | $h_s$         | 7.2   | cm                   |
| Magnets thickness              | $h_M$         | 0.55  | cm                   |
| Air gap (magnet/stator)        | $h_g$         | 0.66  | cm                   |
| Copper maximum temperature     | $T_{cmax}$    | 49    | °C                   |
| Electrical efficiency          | $\eta_{elec}$ | 90    | %                    |
| Magnets maximum magnetic field | $H_{max}$     | 0.596 | MA/m                 |
| Active parts total mass        | $Mass$        | 770.4 | kg                   |
| Active parts total cost        | $Cost$        | 5.04  | k€                   |
| Torque/active parts mass       | $T_{EM}/Mass$ | 62    | Nm/kg                |

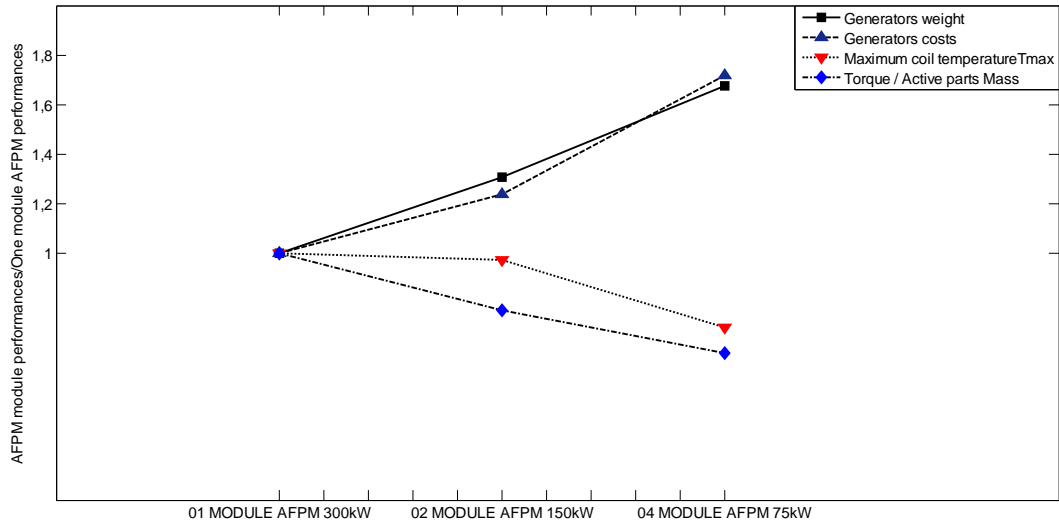


Fig. 5. Modular axial flux generators comparison.

To decrease or even eliminate these ripples, an even number modular arrangement should be considered. In addition, each module is shifted by a specifically mechanical angle (i.e.  $\delta = k\pi/6p$  or an integral windings distribution and  $k$  is an integer). Figure 6 illustrates therefore the electromagnetic torques of a 1-module 300 kW AFPM generator and a 2-module 150kW AFPM generators.

#### IV. Conclusion

This research note has proposed a comparative study of a modular structure of axial flux generators that have been optimized to be inserted in a POD to be directly driven by an MCT. Preliminary results show that, apart from improving the MCT thermal behavior, modular arrangements lead to a significant decrease of the turbine torque ripples if the modules are adequately spatially shifted. However, it has been shown that the modular integration leads to actives parts masses and costs oversizing. In this context, compromises should be carried-out in terms of reliability and fault-tolerance.

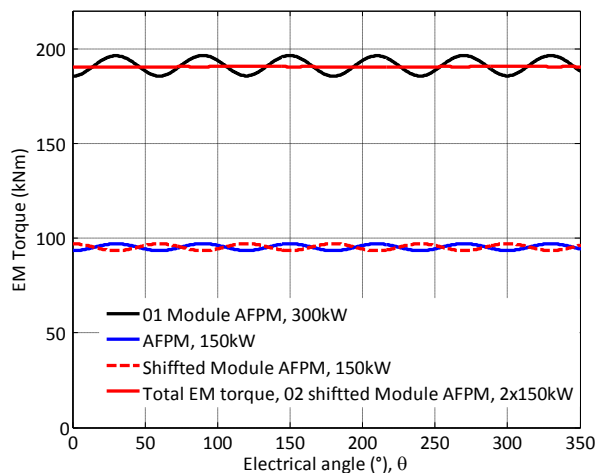


Fig. 6. Electromagnetic torque comparisons.

#### References

- [1] S. Benelghali, M.E.H. Benbouzid and J.F. Charpentier, "Generator systems for marine current turbine applications: A comparative study," *IEEE Journal of Oceanic Engineering*, vol. 37, n°3, pp. 554-563, July 2012.
- [2] S. Benelghali, M.E.H. Benbouzid and J.F. Charpentier, "Marine tidal current electric power generation technology: State of the art and current status," in *Proceedings of the 2007 IEEE IEMDC*, Antalya (Turkey), vol. 2, pp. 1407-1412, May 2007.
- [3] Y. Li and H.K. Florig, "Modeling the operation and maintenance costs of a large scale tidal current turbine farm," in *Proceedings of the IEEE OCEANS'06*, Boston (USA), September 2006.
- [4] R.E. Harris, L. Johanning and J. Wolfram, "Mooring systems for wave energy converters: A review of design issues and choices," in *Proceedings of the 2004 MAREC*, Blyth (UK), pp. 1-10, July 2004.
- [5] R.S. Semken, M. Polikarpova, P. Roytta, J. Alexandrova, J. Pyrhonen, J. Nerg, A. Mikkola and J. Backman, "Direct-drive permanent magnet generators for high-power wind turbines: Benefits and limiting factors," *IET Renewable Power Generation*, vol. 6, n°1, pp. 1-8, January 2012.
- [6] M. Joorabian, A.Z. Nejad, and K. Malekian, "Practical aspects of constructing a wind Ironless axial flux direct-drive generator," *International Review of Electrical Engineering*, vol. 4, n°2, pp. 235-241, March-April 2009.
- [7] U. Kurt, G. Onbilgin and O. Ozgonenel, "Defining design criteria of an axial flux permanent magnet synchronous generator," *International Review of Electrical Engineering*, vol. 7, n°1, pp. 3290-3296, February 2012.
- [8] O. Keysan, A.S. McDonald and M. Mueller, "A direct drive permanent magnet generator design for a tidal current turbine (SeaGen)," in *Proceedings of the 2011 IEEE IEMDC*, Niagara Falls (Canada), pp. 224-229, May 2011.
- [9] C. Boccaletti, P. Di Felice, L. Petrucci, E. Santini, "Parametric analysis of axial flux wind generators focused on total harmonic distortion evaluation," *IET Renewable Power Generation*, vol. 5, n°2, pp. 148-159, 2011.
- [10] S. Djebbari, J.F. Charpentier, F. Sculler, M.E.H. Benbouzid and S. Guemard, "Rough design of a double-stator axial flux permanent magnet generator for a rim-driven marine current turbine," in *Proceedings of the 2012 IEEE ISIE*, Hangzhou (China), pp. 1450-1455, May 2012.
- [11] S. Djebbari, J.F. Charpentier, F. Sculler and M.E.H. Benbouzid, "Axial flux permanent magnet generator with rim driven specifications for marine current turbines," *European Journal of Electrical Engineering*, vol. 16, n°2, pp. 145-176, 2013.
- [12] P. Virtic and B. Stumberger, "Analytical analysis of magnetic field and force calculation in a slotless-type permanent magnet linear synchronous machine; verification with numerical analysis," in *Proceedings of the 2007 IEEE IEMDC*, Antalya (Turkey), vol. 2, pp. 963-968, May 2007.

<sup>1</sup>French Naval Academy, IRENav EA 3634, 29240 Brest Cedex 9, France (e-mail: Sofiane.Djebarri@ecole-navale.fr, Jean-Frederic.Charpentier@ecole-navale.fr, Franck.Scuiller@ecole-navale.fr).

<sup>2</sup>University of Brest, EA 4325 LBMS, Rue de Kergoat, CS 93837, 29238 Brest Cedex 03, France (e-mail: Mohamed.Benbouzid@univ-brest.fr).



**Sofiane Djebarri** was born in Algeria in 1984. He received the B.Sc. and M.Sc. degrees in electrical engineering, from the National Polytechnic School, Algiers, Algeria, and the University of Paris-Sud 11, France, in 2009 and 2010 respectively. He is a Teaching and Research assistant at the French Naval Academy since September 2010.

He is currently pursuing Ph.D. studies on electrical machines design for renewable energy applications in collaboration with the University of Brest.



**Mohamed El Hachemi Benbouzid** was born in Batna, Algeria, in 1968. He received the B.Sc. degree in electrical engineering from the University of Batna, Batna, Algeria, in 1990, the M.Sc. and Ph.D. degrees in electrical and computer engineering from the National Polytechnic Institute of Grenoble, Grenoble, France, in 1991 and 1994, respectively, and the Habilitation à Diriger des Recherches degree from the University of Picardie "Jules Verne," Amiens, France, in 2000.

After receiving the Ph.D. degree, he joined the Professional Institute of Amiens, University of Picardie "Jules Verne," where he was an Associate Professor of electrical and computer engineering. Since September 2004, he has been with the Institut Universitaire de Technologie of Brest, University of Brest, Brest, France, where he is a Professor of electrical engineering. His main research interests and experience include analysis, design, and control of electric machines, variable-speed drives for traction, propulsion, and renewable energy applications, and fault diagnosis of electric machines.

Prof. Benbouzid is an IEEE Senior Member. He is the Editor-in-Chief of the International Journal on Energy Conversion (IRECON). He is also an Associate Editor of the IEEE TRANSACTIONS ON ENERGY CONVERSION, the IEEE TRANSACTIONS ON INDUSTRIAL ELECTRONICS, the IEEE TRANSACTIONS ON SUSTAINABLE ENERGY, and the IEEE TRANSACTIONS ON VEHICULAR TECHNOLOGY. He was an Associate Editor of the IEEE/ASME TRANSACTIONS ON MECHATRONICS from 2006 to 2009.



**Jean Frédéric Charpentier** (M'02) was born in Tananarive, Madagascar, in 1969. He received the M.Sc. and PhD degrees in electrical engineering from the National Polytechnic Institute of Toulouse, Toulouse, France in 1993 and 1996 respectively.

From 1996 to 1997 he was a post doctoral fellow at Laval University, Québec, Canada. From 1997 to 2002 he was an Assistant Professor at the Institut Universitaire de Technologie of Brest, University of Brest, Brest, France. Since 2002, he has been an

Associate Professor in the French Naval Academy in Brest, France. His current research interests include design aspects on electrical machines and drives, electrical naval propulsion systems and marine renewable energy.



**Franck Scuiller** was born in Brest, France, in 1977. He received the electrical engineer degree (M.Sc. degree) from ENSIEG, INPG (Grenoble National Polytechnic Institute) in 2001 and the Ph.D. degree from "Arts et Metiers ParisTech" in 2006.

In 2007, he worked as a lecturer in the French Naval Academy. From 2008 to 2011, he was a Technical Project Manager in warship electric power system for French DCNS company. Since September 2011, he has been an Associate Professor in the French

Naval Academy, Brest, France. His research interesting include marine current turbine and ship propulsion with multi-phase machine, ship grid modeling and simulation, new grid topologies for improving the electrical energy availability and quality.